Field Inhomogeneity Correction based on Gridding Reconstruction for Magnetic Resonance Imaging

H. Eggers, T. Knopp, and D. Potts

Abstract-Spatial variations of the main field give rise to artifacts in magnetic resonance images if disregarded in reconstruction. With non-Cartesian k-space sampling, they often lead to unacceptable blurring. Data from such acquisitions are usually reconstructed with gridding methods and optionally restored with various correction methods. Both types of methods essentially face the same basic problem of adequately approximating an exponential function to enable an efficient processing with Fast Fourier Transforms. Nevertheless, they have commonly addressed it differently so far. In the present work, a unified approach is pursued. The principle behind gridding methods is first generalized to non-equispaced sampling in both domains and then applied to field inhomogeneity correction. Three new iterative algorithms are derived in this way from a straightforward embedding of the data into a higher dimensional space. Their evaluation in simulations and phantom experiments with spiral k-space sampling shows that one of them promises to provide a favorable compromise between fidelity and complexity compared with existing algorithms. Moreover, it allows a simple choice of key parameters involved in approximating an exponential function and a balance between the accuracy of reconstruction and correction.

Index Terms—Magnetic resonance imaging, image reconstruction, gridding, field inhomogeneity, off-resonance correction, conjugate phase reconstruction, iterative reconstruction, spiral imaging

I. INTRODUCTION

Magnetic resonance imaging (MRI) relies on a strong, homogeneous main field. While the field strength determines the net magnetization available for signal generation, the field homogeneity ensures adequate coherence between the precession of individual spins within one voxel and thus sufficient signal lifetime for an efficient detection. More subtle variations of the field strength between different voxels lead to a distortion of the Fourier encoding used to spatially resolve the received signal and, without correction, to artifacts in images. These artifacts are mainly limited to geometric distortion and intensity variation for Cartesian acquisitions, which sample the spatial frequency domain of images, the so-called k-space, on a Cartesian grid. For non-Cartesian acquisitions, however, more severe blurring and other artifacts arise.

Field inhomogeneity is in non-Cartesian imaging usually compensated by either a direct conjugate phase reconstruction (CPR) [1] or an iterative algebraic reconstruction [2]. Both

are closely related, since one iteration of the latter typically involves one application of the former and its adjoint. The individual methods mainly differ in how they make the distortion of the Fourier encoding amenable to a processing with Fast Fourier Transforms (FFTs). They all introduce for this purpose a coarse segmentation, i.e. discretization, in either the time domain [1] or the frequency domain [3] and require for each resulting segment a separate transformation. In the respective other domain, they perform an interpolation to improve accuracy, such as a Hanning interpolation [1] or more sophisticated linear combinations [4], [5].

If field inhomogeneity is disregarded, non-Cartesian acquisitions are commonly reconstructed with gridding methods [6]. These first convolve the non-equispaced k-space samples with a window function of finite extent and resample them to an oversampled Cartesian grid to then employ FFTs for the transformation to image space. They also include a weighting of both the original samples and the resulting images to compensate for variations in sampling density and to counteract the effect of the convolution. Remarkably, they do not attempt to estimate the spatial frequency spectrum of the final images on the Cartesian grid. Gridding methods have been demonstrated to reach a better compromise between accuracy and complexity than simpler methods that rely on an interpolation in k-space only. Alternatively, non-uniform FFTs (NFFTs) have been considered for the transformation of the non-equispaced k-space samples to image space [7].

In fact, the same basic problem of adequately approximating an exponential function to enable an efficient processing with FFTs underlies the field inhomogeneity correction and the reconstruction of non-Cartesian acquisitions. Nevertheless, it has been addressed differently so far. The similarity between both has only recently been realized and exploited to suggest a conceptually very simple approach to field inhomogeneity correction [8]. By embedding the k-space samples and the image pixels into higher dimensional spaces, the processing can essentially be reduced to a Fourier transform. To cope with the irregular sampling in both domains, the use of special NFFTs was advocated.

We start in the present work from this idea and derive less complex and more accurate approaches to field inhomogeneity correction from it, which still adhere to a unified treatment of reconstruction and correction. In the next section, the problem of reconstructing images from non-Cartesian acquisitions in the presence of field inhomogeneity is described mathematically. The approximation that NFFTs are based on is then generalized to irregular sampling in both domains. With its help, three new algorithms for field inhomogeneity correction

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are derived from the proposed embedding of the data into higher dimensional spaces. They are evaluated in simulations and phantom experiments and compared against existing algorithms. Finally, their advantages and disadvantages are discussed.

II. THEORY

In MRI, the demodulated signal $s\left(t\right)$ received from an object with a magnetization $m\left(\boldsymbol{r}\right)$ at a reference time point t=0 is ideally given by

$$s(t) = \int_{\mathbb{D}^3} m(\mathbf{r}) e^{-i\mathbf{k}(t)\cdot\mathbf{r}} d\mathbf{r}.$$
 (1)

 $\boldsymbol{k}(t)$ denotes the trajectory, along which samples are acquired in k-space. It is determined by the time variant gradient field applied during the measurement.

Any inhomogeneity of the main field distorts the Fourier encoding that (1) describes. Taking this imperfection into account, $s\left(t\right)$ is more accurately modelled by

$$s(t) = \int_{\mathbb{D}^3} m(\mathbf{r}) e^{-i\omega(\mathbf{r})t} e^{-i\mathbf{k}(t)\cdot\mathbf{r}} d\mathbf{r}.$$
 (2)

 $\omega(r)$ denotes the angular off-resonance frequency, which is proportional to the local deviation of the field strength. Other imperfections, such as relaxation, are not considered in this work.

We restrict ourselves to 2D imaging from now on. The sampled area of k-space is then confined to $\boldsymbol{k} \in [-\pi,\pi]^2$, and the covered field of view to $\boldsymbol{r} \in [-\frac{N_1}{2},\frac{N_1}{2}] \times [-\frac{N_2}{2},\frac{N_2}{2}]$. Discretizing the integral in (2) on N_1N_2 equispaced voxel positions \boldsymbol{r}_ρ and the signal s(t) on M time points t_κ yields

$$s_{\kappa} \approx \tilde{s}_{\kappa} := \sum_{\rho=0}^{N_1 N_2 - 1} m_{\rho} e^{-i\omega_{\rho} t_{\kappa}} e^{-i\boldsymbol{k}_{\kappa} \cdot \boldsymbol{r}_{\rho}}, \tag{3}$$

where $s_{\kappa}:=s(t_{\kappa}),\ \tilde{s}_{\kappa}:=\tilde{s}(t_{\kappa}),\ m_{\rho}:=m(\boldsymbol{r}_{\rho}),\ \omega_{\rho}:=\omega(\boldsymbol{r}_{\rho}),$ and $\boldsymbol{k}_{\kappa}:=\boldsymbol{k}(t_{\kappa})$. Using the vectors

$$s := (s_{\kappa})_{\kappa=0, M-1}$$

$$\mathbf{m} := (m_{\rho})_{\rho=0,\dots,N_1N_2-1},$$

and the matrix

$$\boldsymbol{H} := \left(\mathrm{e}^{-\mathrm{i}\omega_{\rho}t_{\kappa}} \mathrm{e}^{-\mathrm{i}\boldsymbol{k}_{\kappa}\cdot\boldsymbol{r}_{\rho}} \right)_{\kappa=0,\dots,M-1;\; \rho=0,\dots,N_{1}N_{2}-1} \,,$$

this may be rewritten as

$$spprox Hm$$
 .

We propose determining m by a weighted least squares approach

$$\|\mathbf{s} - \mathbf{H}\mathbf{m}\|_{\mathbf{W}} = \sqrt{\sum_{\kappa=0}^{M-1} w_{\kappa} |s_{\kappa} - \tilde{s}_{\kappa}|^2} \stackrel{\mathbf{m}}{\to} \min, \quad (4)$$

with factors $w_\kappa>0$ that compensate for variations in the local sampling density. It leads to the weighted normal equation of first kind

$$\boldsymbol{H}^{\mathsf{H}}\boldsymbol{W}\boldsymbol{H}\boldsymbol{m} = \boldsymbol{H}^{\mathsf{H}}\boldsymbol{W}\boldsymbol{s}\,,\tag{5}$$

where $W:=\operatorname{diag}(w_\kappa)_{\kappa=0,\dots,M-1}$. Due to the size of this linear system, we suggest solving it iteratively with a suitable variant of the conjugate gradient method, such as the Conjugate Gradient Normal Equation Residual (CGNR) method. In this way, (4) is in each iteration minimized over a certain Krylov space. Moreover, by choosing a zero vector as initial estimate of m, the intermediate result after one iteration is, except for a scaling factor, identical to the right-hand side of this linear system and thus to the result of the CPR, which is one reason for including W in (5).

The computational complexity of determining m then depends primarily on two factors, the required number of iterations and the required effort per iteration. The first factor is mostly linked to the employed initial estimate of m and to the condition of H. We stick with a zero vector as initial estimate and apply no additional preconditioning in this work, yet the presence of W in (5) is also motivated by its beneficial effect on the condition of the system matrix. The second factor is mainly influenced by the multiplication of a vector with the matrix H or H^H . Due to the distortion of the Fourier encoding, this product may not simply be implemented with FFTs. It may efficiently be realized with NFFTs, however, as we show in the remainder of this section.

Following [8], we start by embedding the data in both domains in a higher dimensional space. For this purpose, we set $\mathbf{k}'_{\kappa} := ((\mathbf{k}_{\kappa})^{\top}, t_{\kappa})^{\top}$ and $\mathbf{r}'_{\rho} := ((\mathbf{r}_{\rho})^{\top}, \omega_{\rho})^{\top}$, i.e. we add a time dimension to the spatial frequency domain and an off-resonance frequency dimension to the spatial domain. (3) may then be rewritten as

$$s_{\kappa} \approx \sum_{\rho=0}^{N_1 N_2 - 1} m_{\rho} e^{-i \mathbf{k}_{\kappa}' \cdot \mathbf{r}_{\rho}'}$$
 (6)

The samples in both domains are now non-equispaced.

The NFFT and its adjoint, which are also known as non-uniform FFT of type 1 and type 2, are summarized in App. A. They require the samples in one domain to be equispaced. Unlike these standard NFFTs, a so-called NNFFT, or non-uniform FFT of type 3, was first suggested in [9] and later treated in more detail in [10], which permits the fast calculation of the Fourier transform of a vector of non-equispaced samples at a vector of non-equispaced positions. It basically constitutes a combination of the NFFT and its adjoint. Applying it to the computation of (6), an approach we will call 3D NNFFT, entails a 3D FFT and in both domains a 3D local linear combination. Additionally, it demands two multiplicative oversampling factors, which increase the length of the Fourier transform for all three dimensions. The evaluation of the adjoint of (6), i.e. of the sum

$$\sum_{\kappa=0}^{M-1} s_{\kappa} e^{\mathbf{i} \mathbf{k}_{\kappa}' \cdot \mathbf{r}_{\rho}'}, \tag{7}$$

involves the same effort.

To use the standard NFFTs instead, we have to resample the data in one domain to a Cartesian grid. Preferably, the spatial domain is chosen, since it involves no effort for r. We employ

the approximation

$$e^{2\pi ikx} \approx \frac{1}{\alpha N \hat{\varphi}(x)} \sum_{l=-\alpha N/2}^{\alpha N/2-1} \psi(k - \frac{l}{\alpha N}) e^{2\pi i \frac{lx}{\alpha N}}$$
 (8)

for this purpose, with a real oversampling factor α , an integer constant N, and the Fourier transform $\hat{\varphi}(x)$ and the truncation $\psi(k)$ of a window function $\varphi(k)$ with a kernel size of 2m, where $k \in [-\frac{1}{2} + \frac{m}{\alpha N}, \frac{1}{2} - \frac{m}{\alpha N}]$ and $x \in [-\frac{N}{2}, \frac{N}{2}]$. It is closely related to the approximation underlying the standard NFFTs (A-4) and derived in App. B. Moreover, we choose an integer constant N_3 such that

$$\frac{\omega_{\rho}t_{\kappa}}{2\pi} \in \left[-\frac{N_3}{4} + \frac{m}{2\alpha}, \frac{N_3}{4} - \frac{m}{2\alpha} \right]$$

for all ρ and κ , and a scaling factor W such that

$$\frac{\omega_\rho}{W} \in [-\frac{1}{2} + \frac{m}{\alpha N_3}, \frac{1}{2} - \frac{m}{\alpha N_3}]$$

for all ρ . As the computational complexity grows with it, N_3 should be kept as small as possible. Centering ω_ρ and t_κ , which involves in both domains a multiplication with a phase, is, therefore, advantageous. We then obtain with (8)

$$\begin{array}{rcl} \mathrm{e}^{-\mathrm{i}\omega_{\rho}t_{\kappa}} & = & \mathrm{e}^{-2\pi\mathrm{i}\frac{\omega_{\rho}}{W}\frac{Wt_{\kappa}}{2\pi}} \\ & \approx & \frac{1}{\alpha N_{3}\hat{\varphi}\left(-\frac{Wt_{\kappa}}{2\pi}\right)} \sum_{l=-\alpha N_{3}/2}^{\alpha N_{3}/2-1} \psi\left(\frac{\omega_{\rho}}{W} - \frac{l}{\alpha N_{3}}\right) \\ & \times \mathrm{e}^{-\mathrm{i}\frac{Wt_{\kappa}l}{\alpha N_{3}}} \end{array}$$

and by insertion in (3)

$$s_{\kappa} \approx \frac{1}{\alpha N_{3} \hat{\varphi} \left(-\frac{Wt_{\kappa}}{2\pi}\right)} \sum_{l=-\alpha N_{3}/2}^{\alpha N_{3}/2-1} \sum_{\rho=0}^{N_{1}N_{2}-1} m_{\rho} \times \psi \left(\frac{\omega_{\rho}}{W} - \frac{l}{\alpha N_{3}}\right) e^{-i\mathbf{k}_{\kappa} \cdot \mathbf{r}_{\rho}} e^{-i\frac{Wt_{\kappa}l}{\alpha N_{3}}}.$$
(9)

Setting $\boldsymbol{k}_{\kappa}'' := ((\boldsymbol{k}_{\kappa})^{\top}, Wt_{\kappa}/(\alpha N_3))^{\top}$ and $\boldsymbol{r}_{(\rho,l)}'' := ((\boldsymbol{r}_{\rho})^{\top}, l)^{\top}$ finally yields

$$s_{\kappa} \approx \frac{1}{\alpha N_{3} \hat{\varphi} \left(-\frac{W t_{\kappa}}{2\pi}\right)} \sum_{l=-\alpha N_{3}/2}^{\alpha N_{3}/2-1} \sum_{\rho=0}^{N_{1} N_{2}-1} m_{\rho}$$
$$\times \psi \left(\frac{\omega_{\rho}}{W} - \frac{l}{\alpha N_{3}}\right) e^{-i \mathbf{k}_{\kappa}^{"} \cdot \mathbf{r}_{(\rho,l)}^{"}}. \tag{10}$$

The matrix-vector product may thus be realized by a 3D NFFT. Hence, we will call this approach 3D NFFT. It requires a 3D FFT and a 3D local linear combination in the spatial frequency domain, and it introduces two multiplicative oversampling factors only for the added third dimension. The adjoint reads

$$\sum_{l=-\alpha N_3/2}^{\alpha N_3/2-1} \psi \left(\frac{\omega_{\rho}}{W} - \frac{l}{\alpha N_3} \right) \sum_{\kappa=0}^{M-1} \frac{s_{\kappa}}{\alpha N_3 \hat{\varphi} \left(\frac{Wt_{\kappa}}{2\pi} \right)} \times e^{i\mathbf{k}_{\kappa}^{\prime\prime} \cdot \mathbf{r}_{(\rho,l)}^{\prime\prime}}$$
(11)

and may also be computed with a 3D NFFT, but additionally with a sparse summation over l.

We now consider separating the 3D domains into $2D \otimes 1D$ domains. In this way, the Fourier transform along the added

third dimension can be replaced by an explicit sum, which appears beneficial in view of the sparseness of the data in the 3D spaces. By merely rearranging (9), we obtain

$$s_{\kappa} \approx \sum_{l=-\alpha N_{3}/2}^{\alpha N_{3}/2-1} \frac{e^{-i\frac{Wt_{\kappa}l}{\alpha N_{3}}}}{\alpha N_{3}\hat{\varphi}\left(-\frac{Wt_{\kappa}}{2\pi}\right)} \sum_{\rho=0}^{N_{1}N_{2}-1} m_{\rho} \times \psi\left(\frac{\omega_{\rho}}{W} - \frac{l}{\alpha N_{3}}\right) e^{-i\mathbf{k}_{\kappa} \cdot \mathbf{r}_{\rho}}.$$
 (12)

With this approximation, the matrix-vector product may be calculated by αN_3 2D NFFTs and a summation over l, an approach we will call 2D \otimes 1D NFFT-F. It involves a 2D FFT and a 2D local linear combination in the spatial frequency domain for each NFFT, and it demands only one oversampling factor for each dimension. The adjoint is given by

$$\sum_{l=-\alpha N_3/2}^{\alpha N_3/2-1} \psi\left(\frac{\omega_{\rho}}{W} - \frac{l}{\alpha N_3}\right) \sum_{\kappa=0}^{M-1} \frac{s_{\kappa} e^{i\frac{Wt_{\kappa}l}{\alpha N_3}}}{\alpha N_3 \hat{\varphi}\left(\frac{Wt_{\kappa}}{2\pi}\right)} e^{i\boldsymbol{k}_{\kappa} \cdot \boldsymbol{r}_{\rho}}, (13)$$

where the summation over l is sparse.

We derive a variant of this approach by defining another scaling factor T such that

$$\frac{t_\kappa}{T} \in [-\frac{1}{2} + \frac{m}{\alpha N_3}, \frac{1}{2} - \frac{m}{\alpha N_3}]$$

for all κ . Using again the approximation (B-1), we get

$$\begin{array}{rcl} \mathrm{e}^{-\mathrm{i}\omega_{\rho}t_{\kappa}} & = & \mathrm{e}^{-2\pi\mathrm{i}\frac{t_{\kappa}}{T}\frac{\omega_{\rho}T}{2\pi}} \\ & \approx & \frac{1}{\alpha N_{3}\hat{\varphi}\left(-\frac{\omega_{\rho}T}{2\pi}\right)} \sum_{l=-\alpha N_{3}/2}^{\alpha N_{3}/2-1} \psi\left(\frac{t_{\kappa}}{T}-\frac{l}{\alpha N_{3}}\right) \\ & \times \mathrm{e}^{-\mathrm{i}\frac{\omega_{\rho}Tl}{\alpha N_{3}}} \,, \end{array}$$

and insertion into (3) yields

$$s_{\kappa} \approx \sum_{l=-\alpha N_{3}/2}^{\alpha N_{3}/2-1} \psi \left(\frac{t_{\kappa}}{T} - \frac{l}{\alpha N_{3}}\right) \sum_{\rho=0}^{N_{1}N_{2}-1} \frac{m_{\rho} e^{-i\frac{\omega_{\rho}Tl}{\alpha N_{3}}}}{\alpha N_{3}\hat{\varphi}\left(-\frac{\omega_{\rho}T}{2\pi}\right)} \times e^{-i\mathbf{k}_{\kappa}\cdot\mathbf{r}_{\rho}}. \tag{14}$$

In this way, the matrix-vector product may also be computed by αN_3 2D NFFTs, followed by a sparse summation over l, an approach we will call 2D \otimes 1D NFFT-T. The effort for the NFFTs is substantially reduced if they are evaluated only at those k-space positions that actually contribute to the sparse summation. For the adjoint

$$\sum_{l=-\alpha N_3/2}^{\alpha N_3/2-1} \frac{e^{i\frac{\omega_\rho T l}{\alpha N_3}}}{\alpha N_3 \hat{\varphi}\left(\frac{\omega_\rho T}{2\pi}\right)} \sum_{\kappa=0}^{M-1} s_\kappa \,\psi\left(\frac{t_\kappa}{T} - \frac{l}{\alpha N_3}\right) \times e^{i\mathbf{k}_\kappa \cdot \mathbf{r}_\rho},\tag{15}$$

the summation over l is no longer sparse, but the effort for the NFFTs may be decreased similarly.

With respect to computational complexity, we conclude that the $2D \otimes 1D$ NFFT-T approach is the most efficient. It combines the advantages of an explicit summation and of an implicit local linear combination along the added third dimension in the spatial frequency domain. It is followed by the $2D \otimes 1D$ NFFT-F and the 3D NFFT approaches, which each

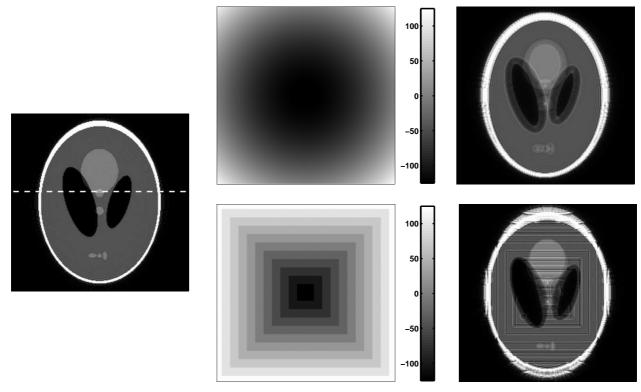


Fig. 1. Image and field maps used in simulations. Shown are a filtered Shepp-Logan phantom on the left, a continuous, parabolic and a discrete, linear field map with off-resonance frequencies in the range of -125 Hz to +125 Hz in the middle, and results of a standard gridding reconstruction on the right. The latter were obtained from simulations of spiral k-space sampling with a readout duration of 32 ms. The dashed line superimposed on the phantom indicates the position where cross sections were taken for comparison.

share one of the advantages with it. The 3D NNFFT approach is the least efficient, since it increases not only the number of segments from αN_3 to $\alpha^2 N_3$, like the 3D NFFT approach, but also the length of the Fourier transform for the other two dimensions from αN_1 x αN_2 to $\alpha^2 N_1 + 2\alpha m$ x $\alpha^2 N_2 + 2\alpha m$.

III. METHODS

We integrated the four outlined gridding-based approaches into the CGNR method to solve (5). We then assessed them in simulations and phantom experiments using spiral k-space sampling. By stopping the CGNR method after either one or several iterations, we covered the cases of a direct CPR and an iterative algebraic reconstruction.

A. Simulations

As reference image, a Shepp-Logan phantom with a resolution of 256 x 256 was employed. We applied a slightly smoothed circular shutter with a radius of $\frac{7}{8}\pi$ to it in k-space, since spiral acquisitions sample only a circular area in k-space and field inhomogeneity seemingly spreads the spatial frequency spectrum over time. The main field was modelled with the two field maps displayed in Fig. 1. The first reflects a continuous, parabolic variation of the field strength, the second a discrete, linear one. The rather unnatural discrete field map was chosen to study how the algorithms cope with a strong local variation of the field strength. Both maps span the same range of off-resonance frequencies of -125 Hz to +125 Hz.

k-space data were calculated by a direct evaluation of (3). We segmented the acquisition into 12 spiral interleaves with 13332 samples each, including a twofold oversampling. Two images obtained with a standard gridding reconstruction of these data, i.e. without field inhomogeneity correction, are presented in Fig. 1. An analytical function described in [11] served the sampling density compensation in this and all other cases presented in this work. Obviously, both field distributions give rise to blurring, but a strong local variation of the field strength causes further major artifacts.

The implementation of the 3D NNFFT and 3D NFFT approaches basically involved the interfacing to 3D NNFFT and 3D NFFT algorithms only. The 3D NFFT approach additionally requires a postweighting of the data. The implementation of all 2D approaches was based on the general framework outlined in [12]. It decomposes the approximation of (3) into three steps: A preweighting of the data, a transformation of each segment, and a linear combination. Mathematically, it may be described by

$$m'_{\rho,l} := v_{\rho,l} m_{\rho},$$

$$s'_{\kappa,l} := \sum_{\rho=0}^{N_1 N_2 - 1} m'_{\rho,l} e^{-i\mathbf{k}_{\kappa} \cdot \mathbf{r}_{\rho}},$$

$$s_{\kappa} \approx \sum_{l=-\alpha N_3/2}^{\alpha N_3/2 - 1} w_{\kappa,l} s'_{\kappa,l}.$$

For the 2D⊗1D NFFT-T approach, for instance, the weights

 $v_{\rho,l}$ and $w_{\kappa,l}$ are given by

$$v_{\rho,l} = \frac{\mathrm{e}^{-\mathrm{i}\frac{\omega_{\rho}'T^{l}}{\alpha N_{3}} - \mathrm{i}\omega_{\rho}t_{c}}}{\alpha N_{3}\hat{\varphi}\left(-\frac{\omega_{\rho}'T}{2\pi}\right)},$$

$$w_{\kappa,l} = \psi\left(\frac{t_{\kappa}'}{T} - \frac{l}{\alpha N_{3}}\right)\mathrm{e}^{-\mathrm{i}\omega_{c}t_{\kappa}'},$$

where ω_c and t_c denote the central and ω' and t' the centered angular off-resonance frequencies and time points, respectively. In the second step, only those $s'_{\kappa,l}$ need to be computed with the αN_3 NFFTs for which $w_{\kappa,l} \neq 0$. For all approaches, pruning techniques were applied to the FFTs that underlie the NFFTs to improve efficiency in view of the employed oversampling.

Besides these gridding-based approaches, a number of existing 2D algorithms were implemented for comparison. These included a nearest neighbor interpolation with frequency segmentation [3], a Hanning interpolation with time segmentation [1], the more sophisticated Man interpolation with frequency segmentation [4], and a least squares interpolation with time segmentation [5]. Additionally, the counterparts with the respective other segmentation were implemented, exploiting the duality between time and frequency segmentation [12]. For the Man interpolation with time segmentation, a manual variation of the oversampling factor was performed, since no explicit rule for its choice was known. For the least squares interpolation, a Householder transformation was employed to solve the minimization problems.

All algorithms were programmed both in Matlab and in C. They were tested on a conventional workstation equipped with an Intel Xeon processor running at 2 GHz and with 256 MB of memory. The software configuration used was Linux 2.4.21, FFTW 3.0.1, and NFFT 2.0. The latter is available from [13] and essentially differs from a standard gridding reconstruction in two respects only: Instead of a Kaiser-Bessel window, its Fourier transform is employed as window function, and its shape parameter is well defined. Images obtained with these algorithms were assessed both visually and quantitatively. For the latter, we used as measure the root of the sum of squares of the differences between reconstructed and original image pixels, divided by the root of the sum of squares of the original image pixels.

B. Experiments

The experiments were performed on a 1.5 T Achieva whole-body scanner (Philips Medical Systems, Best, The Netherlands). Transversal cross sections of standard imaging phantoms were acquired with a resolution of 256 x 256 pixels using a segmented spiral gradient echo sequence. A field of view of 250 mm, a slice thickness of 10 mm, a flip angle of 90°, a TE of 2 ms, and a TR of 1 s were chosen. The readout duration, i.e. the length of time that data are acquired after each excitation, varied between 10 ms and 60 ms, and the number of spiral interleaves changed accordingly.

Field maps were obtained from two separate measurements, which usually differed in TE by 1 ms. Two images were reconstructed from them and thresholded based on signal intensity.

Their phases were then subtracted and the differences scaled and slightly filtered. To reduce edge effects, the resulting field maps were additionally extrapolated to areas masked out before

IV. RESULTS

A. Simulations

The four gridding-based correction algorithms are analyzed regarding their accuracy in Tab. I. The presented simulation results were obtained with the continuous field map and the settings $\alpha=1.25,\,m=2,\,$ and $\alpha N_3=14.$ After one iteration, all four algorithms yield similar errors, i.e. they achieve a comparable accuracy for the case of a direct CPR. After three iterations, the 3D NNFFT approach produces an about 100% and the 2D \otimes 1D NFFT-F approach an about 20% higher error than the two others. Beyond three iterations, errors did not decrease significantly anymore for any algorithm. These simulation results were found to be representative.

Iteration	1	2	3
3D NNFFT	$5.41 \cdot 10^{-2}$	$1.15 \cdot 10^{-2}$	$1.13 \cdot 10^{-2}$
3D NFFT	$5.31 \cdot 10^{-2}$	$5.83 \cdot 10^{-3}$	$5.42\cdot10^{-3}$
2D⊗1D NFFT-F	$5.39 \cdot 10^{-2}$	$6.99 \cdot 10^{-3}$	$6.39 \cdot 10^{-3}$
2D⊗1D NFFT-T	$5.32 \cdot 10^{-2}$	$5.50\cdot10^{-3}$	$5.21\cdot10^{-3}$

TABLE I

Comparison of the accuracy of different gridding-based correction algorithms. Listed is the normalized root mean square (RMS) error after 1, 2, and 3 iterations.

The running times of the four gridding-based correction algorithms are compiled in Tab. II. As expected from the theoretical considerations, the 3D NNFFT approach is the slowest and the $2D\!\otimes\!1D$ NFFT-T approach the fastest. Remarkably, the 3D NFFT approach clearly outperforms the $2D\!\otimes\!1D$ NFFT-F approach, indicating that the locality of the linear combination along the added third dimension in the spatial frequency domain is the dominant factor.

Algorithm	Running Time	
3D NNFFT	2000 ms	
3D NFFT	1060 ms	
2D⊗1D NFFT-F	1530 ms	
2D⊗1D NFFT-T	840 ms	

TABLE II

COMPARISON OF THE RUNNING TIMES OF DIFFERENT GRIDDING-BASED CORRECTION ALGORITHMS. LISTED IS THE MEASURED COMPUTATION TIME PER ITERATION.

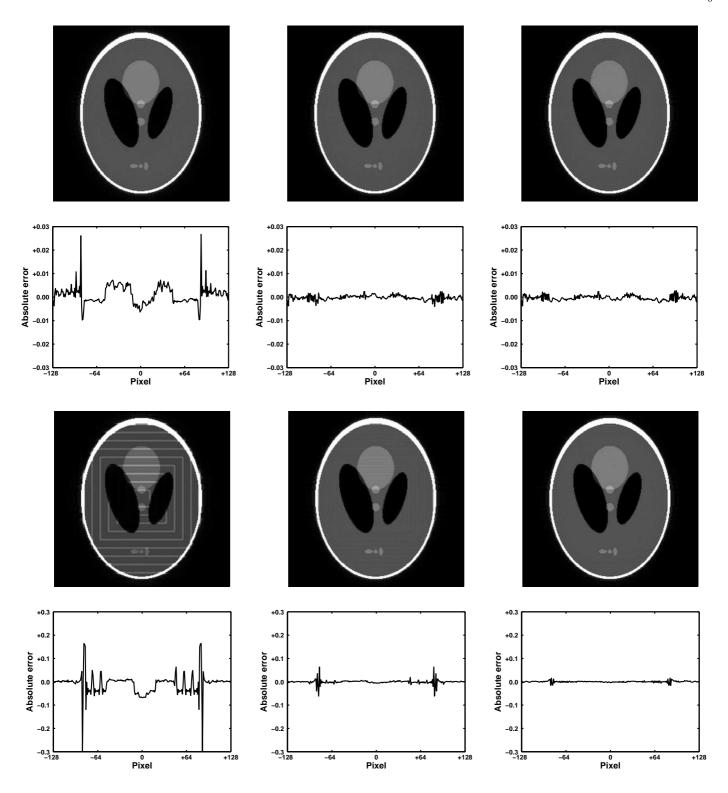
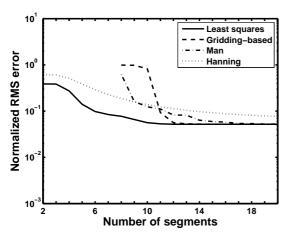


Fig. 2. Results of simulations. Shown are intermediate images after 1, 2, and 3 iterations for the continuous field map at the top and after 1, 2, and 10 iterations for the discrete field map at the bottom. Below them, the differences to the original phantom are plotted for one cross section. The scaling of the graphs varies by one order of magnitude between the top and the bottom.

These two comparisons suggest that the $2D\otimes 1D$ NFFT-T approach is the preferred of the four gridding-based correction algorithms. Its performance is demonstrated with two examples in Fig. 2. For the continuous field map, already the first iteration yields a visually good result. Mainly the

second iteration provides further improvements, in particular at edges. For the discrete field map, the first iteration produces a visually unacceptable result due to artifacts arising from the strong local variation of the field strength. These artifacts are dramatically reduced by the second iteration, but only after



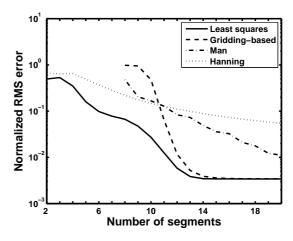


Fig. 3. Comparison of the accuracy of different correction algorithms. Plotted is the normalized root mean square (RMS) error as function of the number of segments in the interpolation after 1 and 3 iterations on the left and on the right, respectively.

about ten iterations the shape of the field map is no longer discernible in the final image. This difference in the speed of convergence is in good agreement with earlier work, which showed that the direct CPR performs reasonably well for spiral k-space sampling only if the field map is smooth [12], [14].

The accuracy of the 2D⊗1D NFFT-T approach is contrasted with that of existing correction algorithms in Fig. 3. The shown simulation results after one and three iterations were again obtained with the continuous field map and with the settings $\alpha=1.25$ and m=2 for the 2D NFFTs. Those for the least squares algorithm were produced with a time segmentation. Using a frequency segmentation instead, errors increased for low number of segments and remained comparable for high number of segments. For the gridding-based approach, the oversampling factor and the kernel size were matched to the respective αN_3 . The approximation becomes senseless below a certain minimum number of segments, which is 8 in this case. The same holds for the Man interpolation, for which results obtained with a frequency interpolation are presented. Slightly lower errors were achieved with a time segmentation, but the manual tuning of the oversampling factor rendered it impracticable. We did not include the nearest neighbor interpolation in the graphs, because it basically failed to yield acceptable results for more than one iteration. While the least squares interpolation obviously achieves the lowest error of all, the gridding-based approach reaches comparable levels very rapidly for sufficiently large oversampling factors and kernel sizes. Using, for instance, the same settings as for the 2D NFFTs, αN_3 is 14. The least squares and the gridding-based algorithms provide similar accuracy in this case. Moreover, the overall accuracy is limited by the 2D NFFTs beyond this point, since errors no longer decrease significantly for higher number of segments. Reasonably good results are achieved with the Man interpolation after one iteration, but a high number of segments is needed to reach an adequate accuracy after three iterations. The Hanning interpolation performs poorest of all shown algorithms, except for very low number of segments.

The running times per iteration of the same algorithms are summarized in Tab. III for a fixed number of segments of 14. Since they do not include the initialization, the differences are mainly due to varying amounts of data that are to be regridded per iteration. These amounts are determined by the number of non-zero linear combination coefficients. The least squares and the Man interpolation show the longest running times, since all coefficients are non-zero. By contrast, only a maximum of 2m+1 out of each αN_3 coefficients are non-zero using the gridding-based approach, and only a maximum of 3 using the Hanning interpolation. Consequently, both exhibit significantly shorter running times.

Algorithm	Running Time	
Least squares	1530 ms	
Gridding-based	840 ms	
Man	1530 ms	
Hanning	710 ms	

TABLE III

COMPARISON OF THE RUNNING TIMES OF DIFFERENT CORRECTION ALGORITHMS. LISTED IS THE MEASURED COMPUTATION TIME PER ITERATION, USING COMPARABLE PARAMETER SETTINGS.

B. Experiments

Representative results of the phantom experiments are summarized in Fig. 4. The off-resonance frequencies cover a range of 210 Hz in this example. Using 12 spiral interleaves and a readout duration of 28.5 ms, the 2D \otimes 1D NFFT-T approach yields an almost perfect image after three iterations. The number of segments was 12 in this case, corresponding to $\alpha=1.33$. Using 6 spiral interleaves and a readout duration of 56.5 ms, it still provides an image of good overall quality, although residual artifacts remain visible, in particular near the circumference and the resolution rods of the phantom. The number of segments was 19 in this case, corresponding to $\alpha=1.26$.

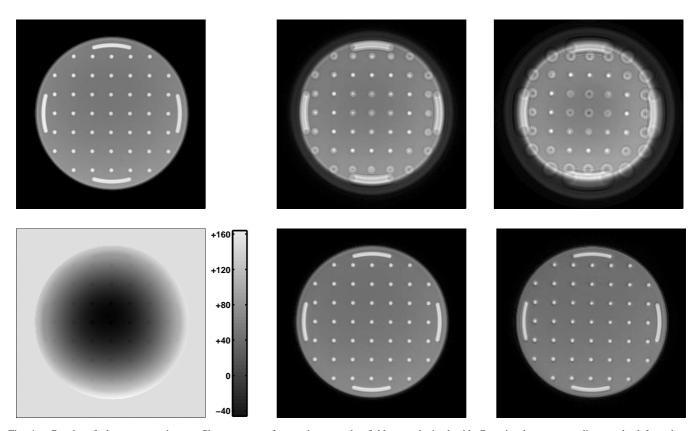


Fig. 4. Results of phantom experiments. Shown are a reference image and a field map obtained with Cartesian k-space sampling on the left, and two corresponding uncorrected and corrected images obtained with spiral k-space sampling in the middle and on the right. The latter were reconstructed from measurements with a readout duration of 28.5 ms and 56.5 ms, respectively.

V. DISCUSSION

The efficient reconstruction of non-Cartesian acquisitions faces the problem of a non-equispaced sampling in the spatial frequency domain. It is adequately solved by both the adjoint NFFT and gridding reconstruction, which rely on the same approximation of an exponential function (A-4). Field inhomogeneity, however, introduces an exponential function with irregular sampling in both domains. We showed that (A-4) can be generalized to this case, leading to the similar approximation (B-1). In this way, the reconstruction and the field inhomogeneity correction of non-Cartesian acquisitions may be founded on the same basic approximation. Based on this concept, we derived three new correction algorithms in the present work.

Among the gridding-based algorithms, the 3D NNFFT approach shows the highest error. We attribute this to the unnecessary interpolation for all spatial dimensions. As already pointed out in [6], gridding remains an approximation even if the samples are equispaced. Therefore, the fact that images are usually discretized on a Cartesian grid should be exploited. The inferior accuracy of the $2D\otimes 1D$ NFFT-F approach compared to the 3D NFFT and $2D\otimes 1D$ NFFT-T approaches is in line with results obtained with the other, existing correction algorithms. Performing the interpolation in the transformed domain of the final result, i.e. in the spatial frequency domain of the images, generally seemed to decrease the artifact level, both visually and quantitatively. We selected the $2D\otimes 1D$ NFFT-T approach for further investigations in this

work for complexity reasons. However, it is worth noting that the 3D NFFT approach achieves a similar accuracy and is conceptually far simpler.

The results summarized in Fig. 3 underline the advantage of combining a local linear combination with a weighting in the transformed domain instead of using a local linear combination only. Beyond a certain number of segments, the accuracy of the Hanning and Man interpolations is substantially inferior to that of the gridding-based approach. Beyond a slightly higher number of segments, which can be calculated explicitly, the accuracy of the 2D NFFTs becomes dominant for the overall accuracy of the gridding-based approach, which then also matches that of the least squares interpolation.

The results compiled in Tab. III highlight the relevance of the amount of data to be regridded. The use of a local neighborhood in the interpolation again appears to be advantageous, as in gridding reconstruction. However, the primary benefit is in this case rather the reduced effort for the 2D NFFTs than for the linear combination. Previously, it has been proposed to eliminate most of the regridding from both the conjugate phase [12] and the algebraic [15] reconstruction. While a detailed comparison remains to be done, the apparent advantage is often offset by either a restriction on the supported k-space trajectories or the requirement of higher oversampling factors to avoid excessive backfolding, as demonstrated for a related problem in parallel imaging [16].

The required number of segments also has a considerable influence on the overall complexity. For the gridding-based

approach, the application of the well-known concept of oversampling permits a simple choice of an adequate number of segments. Moreover, the unified treatment of reconstruction and correction enables a balance of the accuracy of both. Thus, the accuracy of one is not increased beyond the limit set by the other. The same choice may, in principle, be used for the least squares interpolation. Fig. 3 suggests, however, that the error flattens out with a fewer number of segments compared to the gridding-based approach. While this observation may be exploited to further reduce the number of segments especially for the CPR in this particular case, it is not necessarily generalizable.

Another crucial factor is the effort involved in the initialization, which is mainly determined by the calculation of the weights for the linear combination from a given field map. Those approaches which require little such effort are clearly preferable for small number of iterations. The Hanning and the Man interpolation, but also the gridding-based approaches are among these, since they solely need to evaluate a given function. By contrast, the considerably higher initial effort for the least squares interpolation usually pays off for large number of iterations only.

The framework we chose for the reconstruction and field inhomogeneity correction of non-Cartesian acquisitions assumes a piecewise constant field map. While this is usually sufficiently accurate, a more precise model is obtained by also taking intra-voxel gradients into account. These lead to a distortion of the actual k-space trajectory, which varies with spatial position in the image. For a more detailed description of the problem and a potential solution, the reader is referred to [17]. Finally, the inclusion of a weighting in the linear system of equations to be solved entails in theory an SNR penalty [5]. In our experience, however, this penalty was insignificant in practice and was far outweighed by the acceleration of the convergence it provided.

APPENDIX A

In this appendix, the NFFT and its adjoint, i.e. the non-uniform FFT of type 1 and type 2 [9], are outlined for the one dimensional case, and their close relation to gridding reconstruction is highlighted. For a more detailed description, the reader is referred to [18].

Let a function $\varphi \in L^2(\mathbb{R}) \cap L^1(\mathbb{R})$, the so-called window function, be given. Its one periodization

$$\tilde{\varphi}(k) := \sum_{p=-\infty}^{\infty} \varphi(k+p)$$

is assumed to have a uniformly convergent Fourier series. Hence, it may be written as

$$\tilde{\varphi}(k) = \sum_{x=-\infty}^{\infty} c_x(\tilde{\varphi}) e^{2\pi i k x},$$

with Fourier coefficients

$$c_x(\tilde{\varphi}) := \int_{-1/2}^{1/2} \tilde{\varphi}(k) e^{-2\pi i kx} dk, \qquad (A-1)$$

where $x \in \mathbb{Z}$. Substituting k by k - k' in (A-1) yields

$$c_x(\tilde{\varphi}) = \int_{-1/2}^{1/2} \tilde{\varphi}(k - k') e^{-2\pi i(k - k')x} dk', \qquad (A-2)$$

which may be approximated by

$$c_x(\tilde{\varphi}) \approx \frac{1}{\alpha N} \sum_{l=-\alpha N/2}^{\alpha N/2-1} \tilde{\varphi}(k - \frac{l}{\alpha N}) e^{-2\pi i(k - \frac{l}{\alpha N})x}$$
 (A-3)

for $k \in [-\frac{1}{2}, \frac{1}{2}]$ and $x = -\frac{N}{2}, \dots, \frac{N}{2}$. The factor $\alpha > 1$ is commonly referred to as the oversampling factor. For the sake of simplicity, N and αN are assumed to be even. Provided that all $c_x(\tilde{\varphi})$ are non-zero, (A-3) may be rewritten as

$$e^{2\pi i kx} \approx \frac{1}{\alpha N c_x(\tilde{\varphi})} \sum_{l=-\alpha N/2}^{\alpha N/2-1} \tilde{\psi}(k - \frac{l}{\alpha N}) e^{2\pi i \frac{lx}{\alpha N}}, \quad (A-4)$$

where $\tilde{\varphi}$ has been replaced by $\tilde{\psi}$. The latter is the one periodization of a truncation of φ defined by

$$\psi(k) := \left\{ \begin{array}{ll} \varphi(k) & k \in [-\frac{m}{\alpha N}, \frac{m}{\alpha N}] \,, \\ 0 & k \not\in [-\frac{m}{\alpha N}, \frac{m}{\alpha N}] \,. \end{array} \right.$$

The support of ψ is determined by 2m, the so-called kernel size. Typically, $m \in \mathbb{N}$ is chosen such that $m \ll N$. The truncation thus reduces the complexity of the evaluation of the right-hand side of (A-4) considerably.

The NFFT, i.e. the non-uniform FFT of type 2, evaluates the trigonometric polynomial

$$f(k) := \sum_{x=-N/2}^{N/2-1} \hat{f}_x e^{-2\pi i kx}$$
 (A-5)

for N given equispaced samples \hat{f}_x at M given non-equispaced positions $k_j \in [-\frac{1}{2},\frac{1}{2}]$. In matrix-vector notation, it reads

$$f = A\hat{f}, \tag{A-6}$$

with

$$\boldsymbol{f} := (f_j)_{j=0,\dots,M-1} ,$$

$$\hat{\boldsymbol{f}} := \left(\hat{f}_x\right)_{x=-N/2,\dots,N/2-1} ,$$

$$\mathbf{A} := (e^{-2\pi i k_j x})_{j=0,\dots,M-1; x=-N/2,\dots,N/2-1}$$
.

Applying the approximation (A-4) to (A-5) yields

$$f_{j} \approx \sum_{l=-\alpha N/2}^{\alpha N/2-1} \tilde{\psi}(k_{j} - \frac{l}{\alpha N}) \sum_{x=-N/2}^{N/2-1} \frac{\hat{f}_{x}}{\alpha N c_{-x}(\tilde{\varphi})} \times e^{-2\pi i \frac{lx}{\alpha N}}, \tag{A-7}$$

where $f_j:=f(k_j)$. Accordingly, \boldsymbol{A} may be approximated by \boldsymbol{BFD} , where \boldsymbol{D} is a diagonal matrix with entries $d_{x,x}=1/c_{-x}(\tilde{\varphi})$, \boldsymbol{F} an oversampled Fourier matrix, which includes the factor $1/(\alpha N)$, and \boldsymbol{B} a sparse matrix with entries $b_{j,l}=\tilde{\psi}(k_j-l/(\alpha N))$.

The adjoint NFFT, i.e. the non-uniform FFT of type 1, evaluates the sum

$$\sum_{j=0}^{M-1} f_j e^{2\pi i k_j x}$$

for M given non-equispaced samples f_j at N given equispaced positions $x = -\frac{N}{2}, \dots, \frac{N}{2} - 1$. It may be considered as a multiplication of f with $A^H \approx D^H F^H B^H$. As pointed out in [7], [18], gridding reconstruction is simply a fast algorithm for the application of $D^H F^H B^H$ to a vector of non-equispaced samples. Including a sampling density compensation, it performs [6]

- 1) a weighting of the data, i.e. a multiplication with a diagonal matrix W,
- 2) a convolution with a window function and a resampling to an oversampled Cartesian grid, i.e. a multiplication with \boldsymbol{B}^{H} ,
- 3) an inverse FFT, i.e. a multiplication with F^{H} ,
- 4) a deapodization, i.e. a multiplication with D^{H} .

Obviously, the adjoint NFFT and gridding reconstruction rely on the same approximation and involve very similar processing. Moreover, the Kaiser-Bessel window, or its Fourier transform, is a particularly good choice for φ in both cases [18], [19]. More sophisticated approaches based on scaling vectors [20], a minimization of the Frobenius norm of certain error matrices [21], or a min-max interpolation [22] did not prove significantly superior.

APPENDIX B

In this appendix, the approximation underlying the NFFT and its adjoint (A-4) is generalized to real x. Starting from

$$\hat{\varphi}(x) := \int_{-\infty}^{\infty} \varphi(k) e^{-2\pi i kx} dk$$

instead of (A-1) leads to

$$\hat{\varphi}(x) = \int_{-1/2}^{1/2} \sum_{p=-\infty}^{\infty} \varphi(k+p) e^{-2\pi i(k+p)x} dk$$

and, with the same steps as from (A-2) to (A-4), to

$$\mathrm{e}^{2\pi\mathrm{i}kx} \approx \frac{1}{\alpha N \hat{\varphi}(x)} \sum_{l=-\alpha N/2}^{\alpha N/2-1} \sum_{p=-\infty}^{\infty} \psi(k - \frac{l}{\alpha N} + p)$$

$$\times \mathrm{e}^{2\pi\mathrm{i}(\frac{l}{\alpha N} + p)x}$$

for $k\in[-\frac12,\frac12]$ and $x\in[-\frac N2,\frac N2].$ Like (A-4), this approximation may be reduced to

$$e^{2\pi ikx} \approx \frac{1}{\alpha N \hat{\varphi}(x)} \sum_{l=-\alpha N/2}^{\alpha N/2-1} \psi(k - \frac{l}{\alpha N}) e^{2\pi i \frac{lx}{\alpha N}}$$
 (B-1)

for $k \in [-\frac{1}{2} + \frac{m}{\alpha N}, \frac{1}{2} - \frac{m}{\alpha N}]$, since the support of ψ is $[-\frac{m}{\alpha N}, \frac{m}{\alpha N}]$. Consequently, (B-1) is a good approximation if $kx \in [-\frac{N}{4} + \frac{m}{2\alpha}, \frac{N}{4} - \frac{m}{2\alpha}]$. It is worth noting that the further restriction of k may in principle be avoided by explicitly taking the periodization into account. This alternative is not explored in the present work, however.

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